

Control Systems

Continuous linear time-invariant systems (LTI)

"easy" $\begin{cases} \dot{q}(t) = A q(t) + B f(t) \\ y(t) = C q(t) + D f(t) \end{cases}$ $\leftarrow$ inhomogeneous system of ODEs

Continuous linear time variant systems

"harder" $\begin{cases} \dot{q}(t) = A(t) q(t) + B(t) f(t) \\ y(t) = C(t) q(t) + D(t) f(t) \end{cases}$

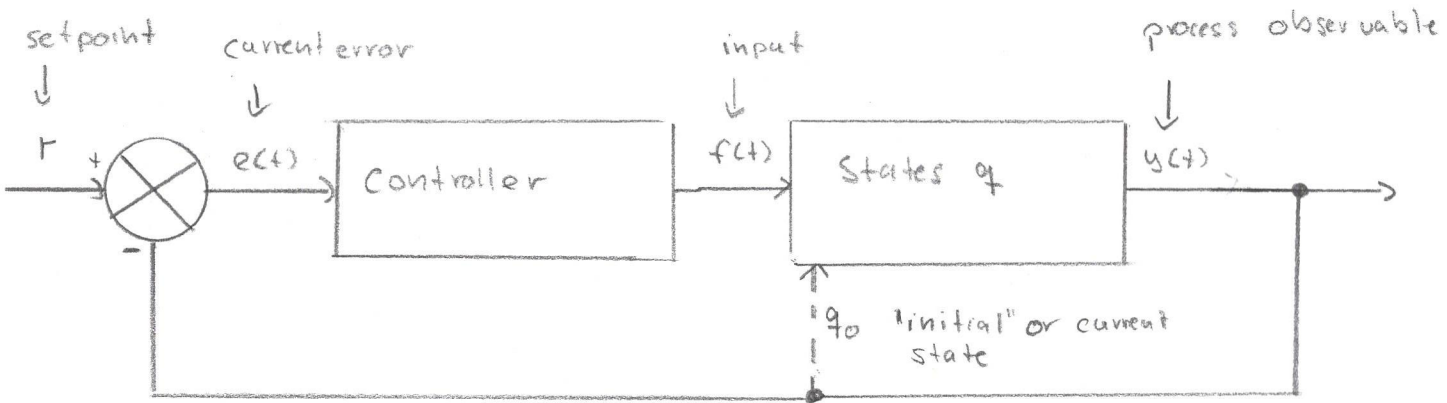
$\uparrow$ linear but coefficients are time dependent

Non-linear systems

$\rightarrow$ no matrix form

"hard" $\begin{cases} \dot{q}_1 = a q_1 + b q_2 + c q_1 q_2 \\ \dot{q}_2 = -d q_1^3 + e \end{cases}$

$$\left. \begin{aligned} \dot{q} &= Aq + Bf \\ y &= Cq + Df \end{aligned} \right\} \text{LTI system}$$

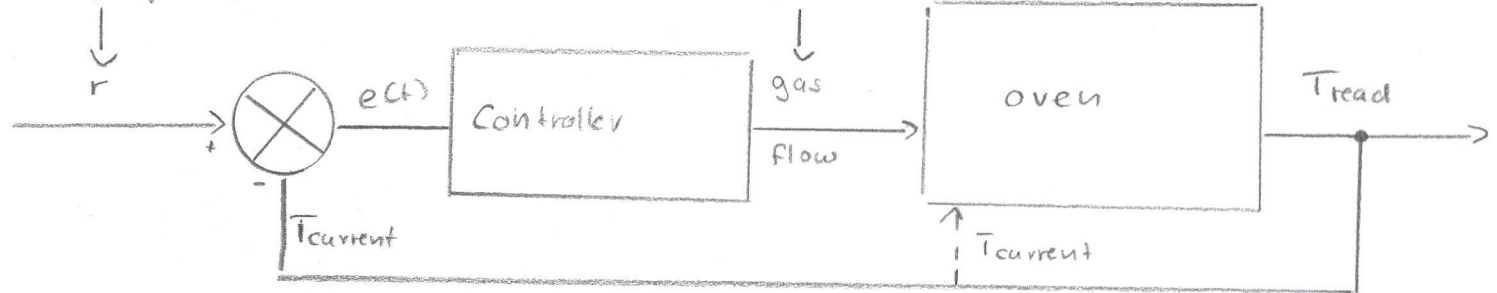


Control problem: how to set input $f(t)$ to drive the system from current state q_0 to desired observable $y(t)$

Example: Oven

MP: Manipulated Variable (value pos/voltage)

SP: Set point



Physical model requires knowledge of V vs. gas flow rate

Discrete Operation ($r = 400$)

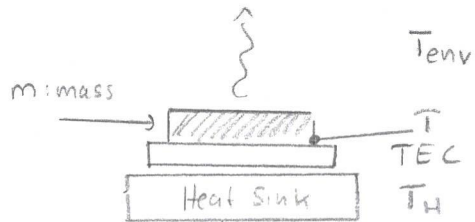
t	T_{read}	$e(t)$	MV
0	150	250	10V
10s	180	220	9V
20s	200	200	8V
30s	210	190	7V

Model (LTI) if available or observe

- know $e(t)$ current
- know $e(t)$ past
- no knowledge of future

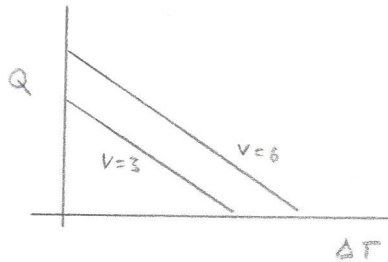
"controller is mapping $e(t)$ to MV"

Concrete Example: Cold Stage



heat flow $[W] = \left[\frac{J}{s} \right]$

$$Q(V, \Delta T) = a_1 \sqrt{V} + a_2 (T_H - T) + a_3$$



How does T change?

$$\frac{dT}{dt} = \underbrace{k(T_{env} - T)}_{\text{Newton's law of cooling}} - \underbrace{\frac{Q}{mc}}_{\text{heat flux + storage}} \quad \left. \vphantom{\frac{dT}{dt}} \right\} \text{physical model of system}$$

Convert model to State space Form

$$\frac{dT}{dt} = kT_{env} - kT - \frac{a_1}{mc} \sqrt{V} + \frac{a_2}{mc} T - \frac{a_3}{mc} + \frac{a_2}{mc} T_H$$

$$\frac{dT}{dt} = \left(-k + \frac{a_2}{mc}\right) T - \frac{a_1}{mc} \sqrt{V} - \frac{a_2}{mc} T_H - \frac{a_3}{mc} + kT_{env}$$

state variable(s): $\vec{q} = T$

inputs: $\vec{f} = V, T_{env}, T_H$

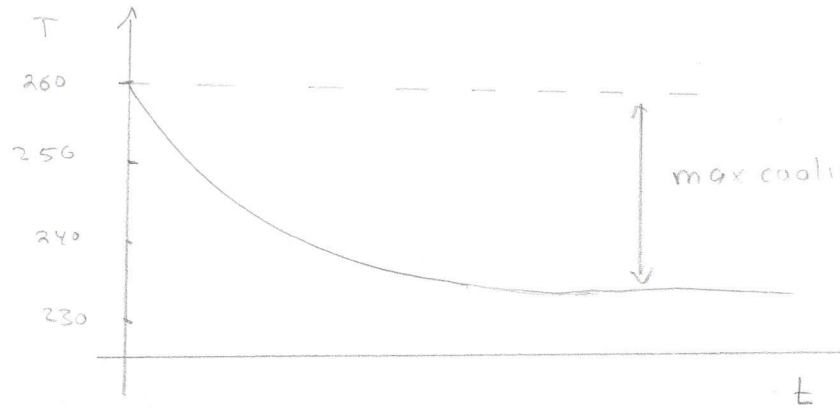
output(s): $\vec{y} = T$

$$q = T \quad q' = T'$$

$$q' = \underbrace{\left[-k + \frac{a_2}{mc}\right]}_A \vec{q} + \underbrace{\left[-\frac{a_1}{mc} \quad -\frac{a_2}{mc} \quad -\frac{a_3}{mc} \quad k\right]}_B \begin{bmatrix} \sqrt{V} \\ T_H \\ 1 \\ T_{env} \end{bmatrix} \vec{f}$$

$$y = \underbrace{[1]}_C \vec{q} + \underbrace{[0 \quad 0 \quad 0 \quad 0]}_D \vec{f}$$

Simple Solution : ODE solver



max cooling depends on V and T_{EC} and T_{env}

function ode(q, p, t)

$A = p[1]$

$B = p[2]$

$f = p[3]$

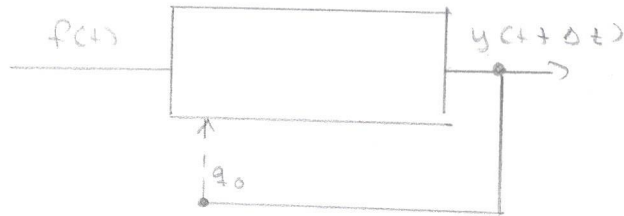
return $Aq + Bf$

end

Discrete Event Simulation

$$q' = Aq + Bf$$

$$y = Cq + Df$$



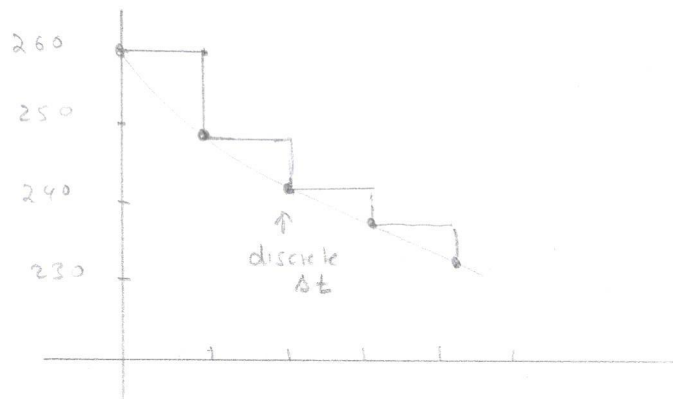
- initialize ODE at $t = t_0$ with IVP q_0, f
 - integrate to $t + \Delta t$
 - evaluate solution $\rightarrow$ build time array, possibly update f
- ← loop

t	q	f
0	⋮	⋮
20	⋮	⋮
40	⋮	⋮
60	⋮	⋮

$\Delta t = 20s$

f may change each Δt

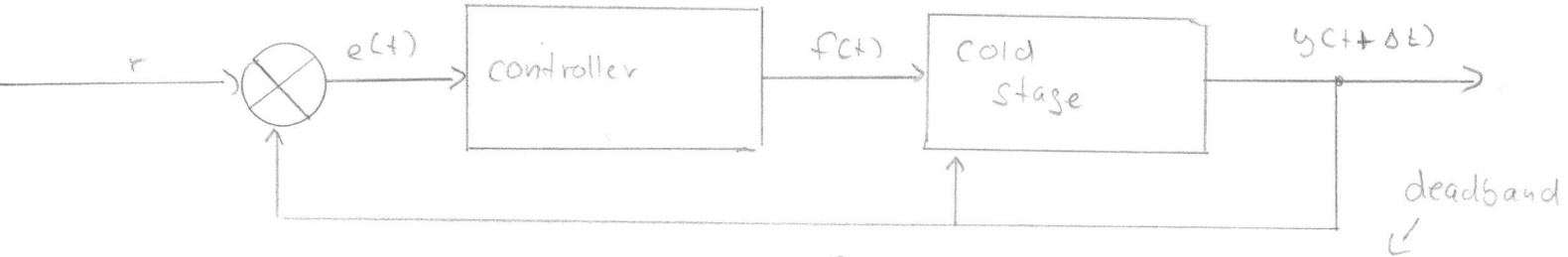
Discrete Event Example



On/Off Control

$r=300$

+1 = cooling $\rightarrow$ TEC = +12V
-1 = heating = -12V



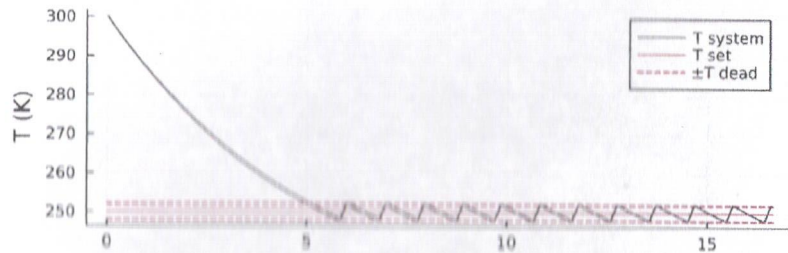
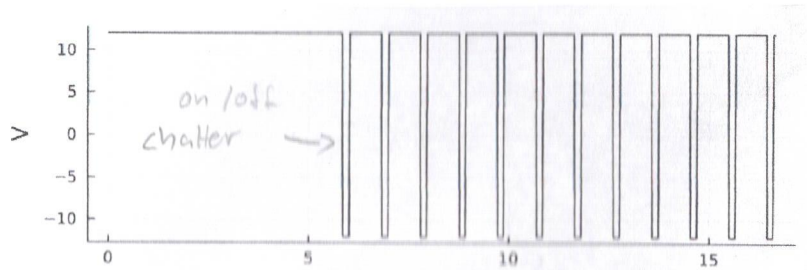
$$e(t) = r - y(t)$$

$$f(t) = \begin{cases} +1 & e(t) + d < 0 \wedge e(t + \Delta t) + d > 0 \\ -1 & e(t) - d > 0 \wedge e(t - \Delta t) - d < 0 \\ \text{no change} & \text{else} \end{cases}$$

past error

t	y(t)	e(t)	e(t)+d	e(t)-d	f(t)
0	303	-3	-1	-5	+1 cooling
1	295	+5	+7	+3	-1 heating
2	299	+1	+3	-1	-1 (no change) "dead band"
3	304	-4	-2	-6	+1 cooling
4	300	0	-1	-5	+1 cooling
5	295	+5	+7	+3	-1 heating
6	299	+1	+3	-1	-1 (no change) "dead band"
7	304	-4	-2	-6	+1 cooling
8	300	0	-1	-5	+1 cooling
9	295	+5	+7	+3	-1 heating
10	299	+1	+3	-1	-1 (no change) "dead band"
11	304	-4	-2	-6	+1 cooling
12	300	0	-1	-5	+1 cooling
13	295	+5	+7	+3	-1 heating
14	299	+1	+3	-1	-1 (no change) "dead band"
15	304	-4	-2	-6	+1 cooling
16	300	0	-1	-5	+1 cooling
17	295	+5	+7	+3	-1 heating
18	299	+1	+3	-1	-1 (no change) "dead band"
19	304	-4	-2	-6	+1 cooling
20	300	0	-1	-5	+1 cooling

Example Cold Stage



} dead band

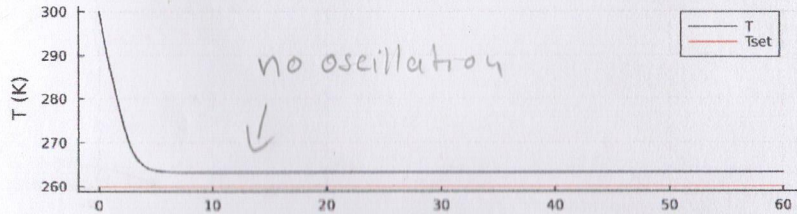
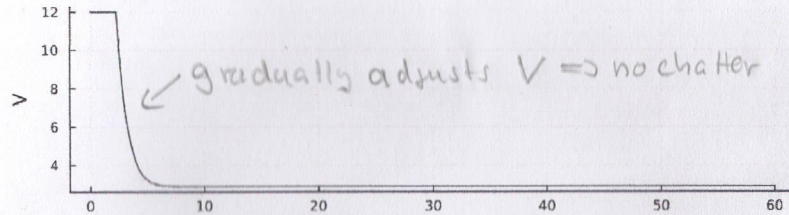
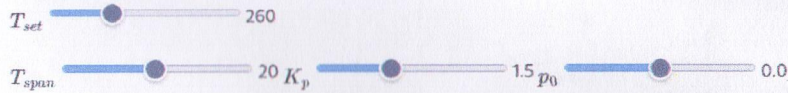
- solution oscillates
- control within $\pm d$
- prone to chatter

Proportional Control

$$g(t) = k_p \frac{\overbrace{r - y(t)}^{\text{error}}}{\underbrace{s}_{\text{span}}} + \underbrace{p_0}_{\text{offset}}$$

$\uparrow$ proportional gain
 $\uparrow$ span
 $\uparrow$ offset

$$f(t) = \begin{cases} -1 & g(t) < -1 \\ 1 & g(t) > 1 \\ g(t) & \text{else} \end{cases}$$



Span: 100% output if not within $\pm s$

} offset error can not be eliminated

Proportional Integral Control (PI)

Integral over past error

$$e(t) = \frac{r - y(t)}{g(t)}$$

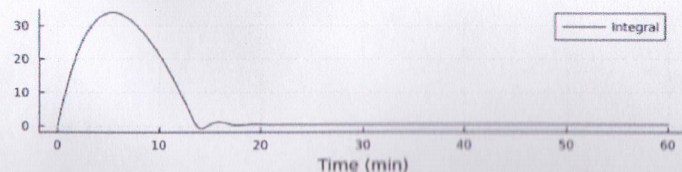
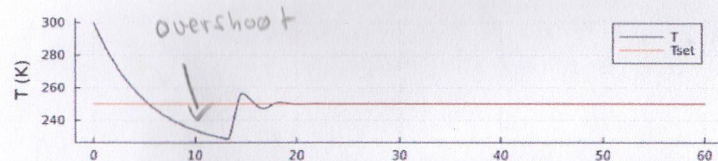
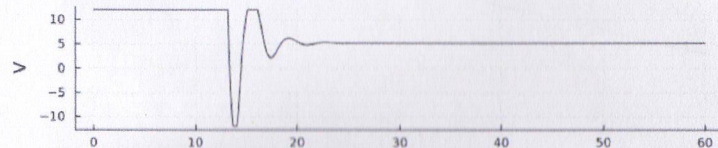
$$g(t) = k_p e(t) + k_i \int_0^t e(\tau) d\tau$$

$$f(t) = \begin{cases} -1 & g(t) < -1 \\ 1 & g(t) > 1 \\ g(t) & \text{else} \end{cases}$$

Solution is sensitive to integral gain

T_{set} 250

T_{span} 20 K_p 1.7 K_i 0.1



- Oscillating solution that generally but not always converges
- no offset error

Proportional - Integral - Derivative (PID) control

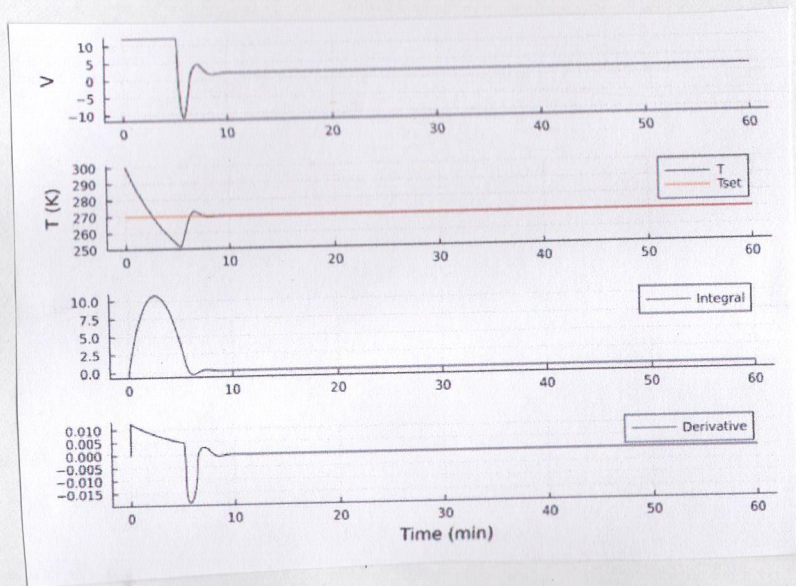
same as PI except

$$g(t) = k_p e(t) + k_i \int_0^t e(\tau) d\tau + k_d \frac{de(t)}{dt}$$

↑
derivative gain

D-term can

- increase convergence rates
- be susceptible to noise amplification

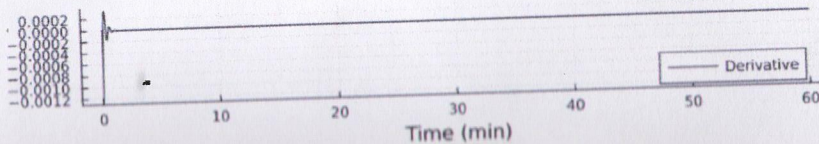
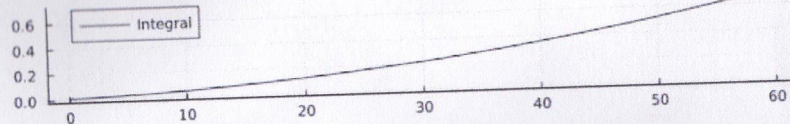
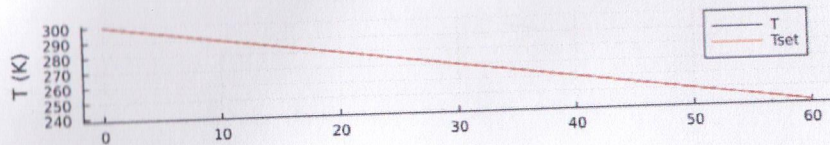
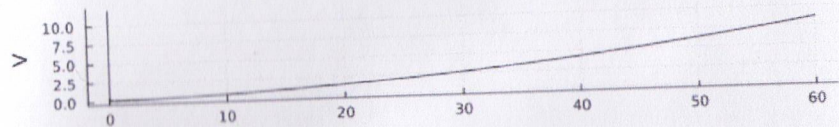


Trajectory Control

$c_r [K \min^{-1}]$

T_{span}

K_p K_i K_d



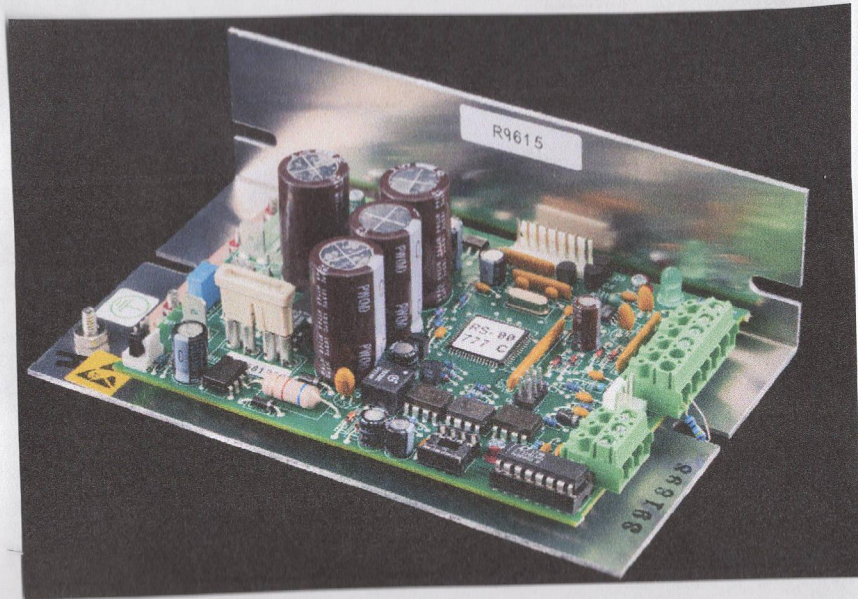
Setpoint r is treated as $r(t)$

e.g. cooling rate $c_r = \frac{dr}{dt} = \text{const}$

- avoids oscillations
- avoids overshoot
- controlled way to steer system

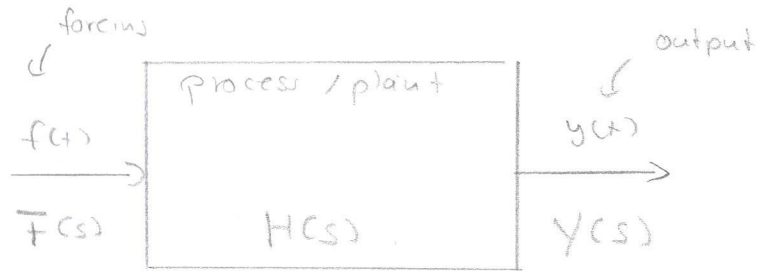
Hardware Controllers (special domain)

PID Temperature Controller



- programmable T_{set} , T_{span} , K_p , K_i , K_d and $span$
- reads temp (thermistor)
- modulates input V to $\pm$ output V for thermoelectrics

Formal Reasoning about control systems



$$\mathcal{L}f'' = s^2 \bar{f} + sf(0) - f'(0)$$

$$\mathcal{L}f' = s\bar{f} - sf(0)$$

$$\mathcal{L}af = a\bar{f}$$

$$f(t) = a_n y^{(n)} + \dots + a_1 y' + a_0 y$$

(zero initial conditions)

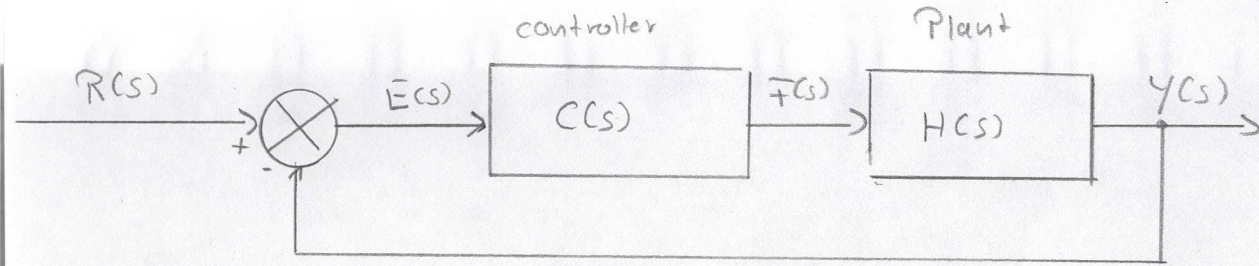
$$\mathcal{L} \bar{f}(s) = (a_n s^n + \dots + a_1 s + a_0) Y(s)$$

$$H(s) = \frac{Y(s)}{\bar{f}(s)} = \frac{1}{a_n s^n + \dots + a_1 s + a_0}$$

↑

transfer function

Closed Loop Control System



$$\frac{Y(s)}{R(s)} = \frac{C(s)H(s)}{1 + C(s)H(s)}$$

↑
system
transfer
function

use this for
stability analysis

Controller

$$f(t) = k_p e(t) + k_i \int_0^t e(\tau) d\tau + k_d \frac{de(t)}{dt}$$

$$C(s) = k_p + \frac{k_i}{s} + k_d s$$

- roots in numerator $\Rightarrow$ zeros $\Rightarrow$ stabilize system
- roots in denominator $\Rightarrow$ poles $\Rightarrow$ destabilize system
- can add noise filter to $C(s)$

$\mathcal{L}^{-1}(Y(s)) \Rightarrow y(t)$: analytical analog to
simulation

Controllability

$$\dot{q} = Aq + Bf$$

→ reach any state $\vec{q}$ using control system $\vec{q}$ is $n \times 1$

$$R = [B \quad AB \quad A^2B \quad \dots \quad A^{n-1}B]$$

↑
controllability matrix

→ system is controllable if R is full row rank, i.e. $\text{rank}(R) = n$

↳ derive through $\mathcal{L} \dot{q}$